

Approximation of ECG Signals Using Chebyshev Nodes and Lagrange-interpolation

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Abstract

An ECG (Electrocardiography) is a graphical representation of electrical behavior of heart and is measured by placing electrodes on specific locations of limbs and chest. The characteristics of these signals change during acquisition and transmission due to addition of noises of variable frequency and amplitude. These noises need to be removed for better clinical evaluation. In this paper, ECG signals of MIT-BIH database have been approximated through a series of steps, i.e., denoising using total variation, segmentation using Bottom Up approach and finally utilizing Chebyshev nodes for Lagrange interpolation method. This paper also explores the concept and the benefit of second difference total variation over the first difference. The performance of the method is analyzed in terms of mean absolute deviation, root mean square deviation, percentage root mean square difference error, signal to noise ratio and cross correlation. The results obtained are found to be better than exiting techniques.

Keywords: ECG signal; Total Variation Denoising; Majorization-Minorization optimization; Chebyshev Nodes; Lagrange Interpolation

1. Introduction

An ECG is an important tool for cardiac related treatments. The signal is acquired through 16 conventional 12-lead ECG in which ten electrodes are placed on the patient's limbs and chest. The resultant signal ^[1, 2] consists of waves viz., P, QRS, T and U of different shapes,

amplitudes and frequencies which are shown in Figure 1.

The characteristics of these signals are changed due to power line interference (PLI), baseline drift, electrode contact noise, motion artifacts, muscle contraction, instrumentation noise etc^[3]. Thus, it is necessary to reduce noises from ECG signals up to the extent where ECG retains diagnostic features^[4].

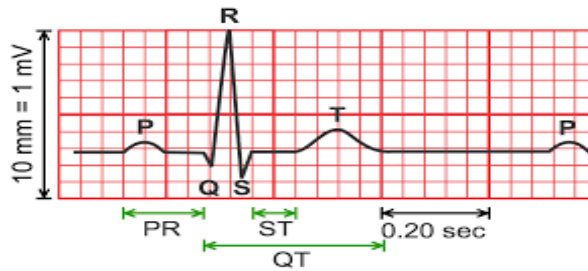


Figure 1: A standard ECG signal with components waves and intervals on ECG grid

Since, noises introduced in ECG signals are of variable frequency and amplitude, significant noise removal cannot be achieved using a single filter. Traditional analog and digital filters were found to suppress ECG components near to 50 Hz frequency. Different types of infinite impulse response filters (IIR) and finite impulse response filters (FIR) with unacceptably long transient time were widely used to reduce PLI noises [5, 6, 7]. Also determination of cut-off frequency for these filters was not so easy. Adaptive filters minimize the error between noisy ECG and a reference ECG with high transient time especially on the QRS complex [8]. Other filters like least mean square (LMS), normalized least mean square (NLMS), transform domain least mean square (TDLMS) which suffered from numerical instability were also found in literature [9]. Baseline wandering is reduced to significant level using linear and polynomial filtering. Low frequency noises were reduced using median filters in [10]. Wavelet transform which allows analysis of signal in both time and frequency scales were also find application in ECG denoising. Various ECG filters using wavelet transform can be found in [11,12,13,14]. Selection of threshold and decomposition level is still a challenge. Neural networks and genetic algorithms were also applied to reduce noise from ECG signals [15]. An effective ECG enhancement technique using total variation was proposed in [16]. In this paper, noises present in ECG signals of MIT-BIH are reduced through first difference and second difference total variation. The effect of regularization parameters on these algorithms is also explored. Characterization of these signals through Lagrange-Chebyshev interpolants is also presented and is found to be more useful.

The rest part of the paper is organized as follows: in section 2, we present the total variation denoising technique based on first and second

difference. In section 3, we give a brief introduction to characterization using Lagrange-Chebyshev interpolants. In section 4, we describe the method and discuss the results obtained. In the last section, we give the conclusions regarding the presented approach.

2. Total variation denoising: First difference-Second difference

The total variation (TV) of N point discrete signal $x(n)$, $1 \leq n \leq N$ is defined as [17]

$$TV(x) = \sum_{n=1}^N |x(n) - x(n-1)| \quad (1)$$

The TV of a signal measures sum of deviations, i.e., difference between consecutive samples of the signal. It is found that signals with high noise have high TV. Therefore, reducing the TV of the signal removes unwanted detail and preserves important diagnostic features [18].

Total variation denoising (TVD) assumes that the noisy data $x(n)$ is of the form

$$x(n) = y(n) + w(n) \quad n = 0, \dots, N \quad (2)$$

where $y(n)$ an approximately piecewise constant signal and $w(n)$ is white Gaussian noise. TVD is defined as an optimization problem which minimizes the cost function (3) for reduction of noises and preservation of sharp edges [19].

$$\arg \min_x \left\{ f(x) = \frac{1}{2} \sum_{n=0}^{N-1} |y(n) - x(n)|^2 + \lambda \sum_{n=1}^N |x(n) - x(n-1)| \right\} \quad (3)$$

The first term in (3) represents the mean square error between the observed and the reconstructed signal, and the second term refers to the total variation. The regularization parameter λ controls the degree of smoothing with larger value for large noise. Increasing λ gives more weight to the TV of the signal [20].

Since ECG signals are not piece-wise constant, TVD using first order differences has the tendency

to introduce a staircase effect leading to small flat regions in the denoised signal. While conventional TVD may be suitable for filtering piecewise constant signals, it is not usually the best denoising method for more general piecewise-smooth signals [21]. For such signals, a higher-order difference can preferably be used instead of the first order differences. So, a variant of TVD is proposed in this paper which would reduce the staircase effect while retaining the quality of the reconstructed signal.

Let the ECG signal x be represented as N point vector

$$x(n) = [x(0), \dots, x(N-1)] \quad (4)$$

Expressing the TV of the signal $x(n)$, in terms of second order differences as

$$TV(x) = \sum_{n=2}^{N-1} \{x(n) - x(n-1)\} - \{x(n-1) - x(n-2)\} \quad (5)$$

Therefore, the optimization problem reduces to

$$\arg \min_x \left\{ f(x) = \frac{1}{2} \sum_{n=0}^{N-1} |y(n) - x(n)|^2 + \lambda \sum_{n=2}^{N-1} |x(n) - x(n-1) - \{x(n-1) - x(n-2)\}| \right\} \quad (6)$$

Our aim is to estimate $y(n)$ from $x(n)$ by minimizing (6). Since l_1 norm is not differentiable, we minimize the objective function using Majorization-Minimization (MM) algorithm.

The MM method solves the optimization problem by replacing a complicated problem by a sequence of simpler problems. Convergence is guaranteed by requiring that the approximating functions majorize the original function at the current solution [22]. If $x, y \in \mathbb{R}$, f and g be real valued functions on \mathbb{R}^n , then the function g majorizes the function f at y if:

- (a) $g(x) \geq f(x)$ for all x
- (b) $g(y) = f(y)$

While minimizing the objective function f iteratively, let $x(k)$ be the current best minimizer at the k^{th} iteration. A majorizing function g is constructed that majorizes f at $x(k)$. If $x(k)$ minimizes g , the procedure is terminated otherwise a new solution $x(k+1)$ is found by minimizing g .

$$f(x^{k+1}) \leq g(x^{k+1}) \leq g(x^k) \quad (7)$$

A new majorizing function is constructed at $x(k+1)$, and the steps are repeated to produce a

decreasing sequence of function values. In order to construct a majorizer for the objective function given by (6), the property of quadratic majorizers has been exploited. The function $f(x) = |x|$ has a quadratic majorizer at each x_k except at $x_k = 0$. If $x(k) \neq 0$ then the majorizer for $f(x)$ is given by [22]

$$|x| = \frac{1}{2|x_k|} x^2 + \frac{1}{2} |x_k| \quad (8)$$

Therefore, instead of minimizing the cost function $f(x)$ directly, the Majorization approach solves a sequence of optimization problems, $G_k(x)$, $k = 0, 1, 2, \dots$, where each function $G_k(x)$, is a majorizer of $f(x)$. Using (8) we can construct the majorizer for (6) as

$$G_k(x) = \frac{1}{2} \sum_n |y(n) - x(n)|^2 + \frac{1}{2} \lambda \sum_n \left[\frac{(x(n) - x(n-1))^2}{|x_k|} + |x_k| \right] \quad (9)$$

Since common ECG contaminants are non-stationary and temporally correlated, time-varying dynamic models are required for the generation of realistic noises.

3. Characterization using Chebyshev- Lagrange interpolation

The Chebyshev polynomials of first kind, degree n are defined [23] as:

$$T_n(x) = \cos(n \cos^{-1}(x)) \text{ for } n \geq 1 \quad (10)$$

The n^{th} degree Chebyshev polynomial has $n + 1$ zeros (nodes or points) in the interval $[-1, 1]$, which can be calculated as:

$$x_j = \cos\left(\frac{2j+1}{2(n+1)}\pi\right) \text{ for } 0 \leq j \leq n \quad (11)$$

The Chebyshev polynomials are orthogonal in the interval $[-1, 1]$ over the weight $w(x) = (1 - x^2)^{-1/2}$. Other properties of Chebyshev polynomials can be found in [24]. Chebyshev interpolation produces a sequence of polynomials $p(x)$ that converge uniformly to $f(x)$ over $[-1, 1]$ [25]. If $f(x)$ is a continuous function on $[-1, 1]$, the polynomial interpolation of degree n can be obtained by interpolating between the values of $f(x)$ at $n + 1$ significant points in the interval. Let $f(x_j)$, $0 \leq j \leq n$ be a set of $N + 1$ numbers representing the samples of ECG sequence vector of length N in $[-1, 1]$. Then there exists a unique

polynomial p of degree $n \leq N$ that interpolates these data, i.e., $p(x_j) \approx f(x_j)$ for each j .

If the interpolating polynomial is

$$p_n(x) = c_0 + c_1x + c_2x^2 + \dots + c_{n-1}x^{n-1} + c_nx^n \quad (12)$$

We require that

$$c_0 + c_1x_j + c_2x_j^2 + \dots + c_{n-1}x_j^{n-1} + c_nx_j^n = f(x_j) \quad (13)$$

In matrix form (15) can be rewritten as

$$\begin{bmatrix} 1 & x_0 & \Lambda & x_0^{n-1} & x_0^n \\ 1 & x_1 & \Lambda & x_1^{n-1} & x_1^n \\ \vdots & \vdots & \vdots & \vdots & \vdots \\ 1 & x_n & \Lambda & x_n^{n-1} & x_n^n \end{bmatrix} \begin{bmatrix} c_0 \\ c_1 \\ \vdots \\ c_n \end{bmatrix} = \begin{bmatrix} f(x_0) \\ f(x_1) \\ \vdots \\ f(x_n) \end{bmatrix} \quad (14)$$

In order for the system (14) to have a unique solution, the Vandermonde determinant on the extreme left should be non singular [26]. The Vandermonde determinant equals the product of the terms $(x_i - x_j)$ for $i > j$, therefore the points x_0, \dots, x_n should be distinct for the determinant to be non zero. Setting the coefficients as the interpolated values $f(x_j)$; $0 \leq j \leq n$, we can write the interpolating polynomial of degree n using Lagrange's formula [27,28] as

$$p_n(x) = \sum_{j=0}^n l_j^n(x) f(x_j) \quad (15)$$

where l_j^n are $(n+1)$ Lagrange polynomials of degree $\leq n$.

$$l_j^n(x) = \prod_{j=0, j \neq i}^n \frac{x - x_j}{x_i - x_j} \quad (16)$$

If $f(x)$ is $n+1$ times continuously differentiable in $[-1, 1]$, then the interpolation error $E_n(x)$ is [29]

$$f(x) - p_n(x) = \frac{1}{n+1} f^{(n+1)}(\xi) \prod_{j=0}^n (x - x_j), \quad \xi \in [-1, 1] \quad (17)$$

Therefore,

$$|E_n(x)| \leq \frac{1}{(n+1)!} \left| \prod_{j=0}^n (x - x_j) \right| \max_{-1 \leq \xi \leq 1} |f^{(n+1)}(\xi)| \quad (18)$$

To minimize the upper bound for $|E_n(x)|$, we can minimize the product $\prod_{j=0}^n (x - x_j)$

If we are free to choose the interpolating points x_0, \dots, x_n within this interval, then the product $\prod_{j=0}^n (x - x_j)$ can be minimized. A better choice of interpolating points x_0, \dots, x_n to ensure uniform convergence is the set of zeros of the Chebyshev

polynomial $T_{n+1}(x)$, instead of equally spaced nodes [27]. The following theorem gives an estimate of the error for the above case.

Theorem 1: Assume that $p_n(x)$ is the Lagrange polynomial that interpolates $f(x)$ at x_0, \dots, x_n . Also assume that these $n+1$ interpolation points are the $(n+1)$ roots of the Chebyshev polynomial $T_{n+1}(x)$, given by (11). Then $\forall x \in [-1, 1]$

$$|f(x) - p_n(x)| \leq \frac{1}{2^n(n+1)!} \max_{-1 \leq \xi \leq 1} |f^{(n+1)}(\xi)| \quad (19)$$

If the interpolation interval for the function $f(x)$ is $x \in [a, b]$, we transform the interval $y \in [-1, 1]$ using

$$x = \frac{(b-a)y + (a+b)}{2} \quad (20)$$

This converts the interpolation problem for $f(x)$ on $[a, b]$ into interpolation problem for $f(x) = g(y)$ in $y \in [-1, 1]$. The Chebyshev points in the interval $y \in [-1, 1]$ are the roots of the Chebyshev polynomial $T_n(y)$, i.e.,

$$y_j = \cos\left(\frac{2j+1}{2(n+1)}\pi\right) \quad 0 \leq j \leq n \quad (21)$$

The corresponding $n+1$ interpolation points in the interval $[a, b]$ using (21) are now

$$x_j = \frac{(b-a)y + (a+b)}{2} \quad 0 \leq j \leq n \quad (22)$$

The interpolation error now is given by

$$|f(x) - p_n(x)| = \frac{1}{2^n(n+1)!} \left| \frac{b-a}{2} \right|^{n+1} \max_{-1 \leq \xi \leq 1} |f^{(n+1)}(\xi)| \quad (23)$$

where $p_n(x)$ is the Lagrange interpolating polynomial based on Chebyshev nodes. Delving deeper into the advantages of using Chebyshev interpolating nodes, we observe that Runge phenomenon does not occur with the effect that the error tends to decrease with the increasing degree of the Lagrange Chebyshev interpolating polynomial, whereas the same may not be true for equally spaced nodes. Alternatively, we can express the n th degree interpolating polynomial $p_n(x)$ as a sum of Chebyshev polynomials $T_k(x_j)$ [30].

$$p_n(x) = \sum_{k=0}^n c_k T_k(x), \quad x \in [-1, 1] \quad (24)$$

where the coefficients c_k are defined as

$$c_k = \frac{2}{n+1} \sum_{j=0}^n f(x_j) T_k(x_j), \quad k = 0, \dots, n \quad (25)$$

where

$$x_j = \cos\left(\frac{(2j+1)\pi}{2(n+1)}\right) \quad j=0, \dots, n \quad (26)$$

Let

$$x = \theta, \theta \in [-\pi, \pi] \quad (27)$$

then,

$$c_k = \frac{2}{n+1} \sum_{j=0}^n g(\theta_j) \cos(k\theta_j) \quad k = 1, \dots, n \quad (28)$$

with

$$\theta_j = \frac{(2j+1)\pi}{2(n+1)} \quad (29)$$

Replacing $f(\cos \theta)$ by a periodic function $g(\theta)$,

$$c_k = \frac{2}{n+1} \sum_{j=0}^n g(\theta_j) \cos(k\theta_j), \quad k = 0, \dots, n \quad (30)$$

Thus c_k is discrete approximation to the Fourier series coefficients

$$c_i^s \approx \frac{1}{\pi} \int_{-\pi}^{\pi} g(\theta) \cos(k\theta) d\theta \quad (31)$$

Applying the trapezoidal rule approximation

$$c_i^s = \frac{1}{\pi} \frac{\pi}{n} \sum_{k=0}^{n+1} g\left(\frac{(2j+1)\pi}{2(n+1)}\right) \cos\left(\frac{(2j+1)\pi}{2(n+1)}\right) \quad (32)$$

which is same as (30) for c_k .

Thus (32) is equivalent to the discrete Fourier transform of the transformed function $g(\theta) = f(\cos \theta)$. Chebyshev interpolation can effectively be considered as the partial sum of the approximate Chebyshev series expansion obtained by replacing the Fourier transform c_k^s by discrete Fourier transform c_k .

4. Methods and Results

Presence of noises degrade the signal quality and changes the morphology of the ECG signal, thus establishing noise removal is an essential step in ECG preprocessing for better performance, analysis and diagnosis. In this paper, first difference, second difference and Lagrange-Chebyshev interpolants are used for noise removal from ECG signals. To test and to compare the performance of the algorithms, we perform all the algorithms in Matlab environment. The signals used

for analysis are taken from MIT-BIH ^[31] arrhythmia database. Each record consists of two channels of signals, which are of 10 seconds duration with 3600 samples, sampled at a rate of 360 Hz with 11 bits per sample of resolution.

4.1 Computational Performance

Let $x(n)$ be the original signal and $y(n)$ be reconstructed ECG signal of length N . The error between the signals is evaluated as:

$$e(n) = x(n) - y(n)$$

The performance of algorithms with respect to the signals is evaluated in terms of following parameters:

4.1.1 Mean Absolute Deviation

Mean absolute deviation (MAD) provides average of absolute deviation of reconstructed signal from the original signal.

$$MAD = \frac{1}{N} \sum_{i=1}^N |e(n)|$$

4.1.2 Root Mean Square Difference

The root-mean-square difference (RMSD) measures mean of the differences between reconstructed and the original signal ^[32]. The RMSD is more useful when large errors are particularly undesirable

$$MSD = \sqrt{\frac{\sum_{i=1}^N |e(n)|^2}{N}}$$

4.1.3 Percentage Root-Mean-Square Difference

The percentage root mean square difference ^[32] (PRD) is calculated by:

$$PRD = \sqrt{\frac{\sum_{i=1}^N |e(n)|^2}{\sum_{i=1}^N |x(n) - \bar{x}|^2}}$$

where \bar{x} is the mean of the original signal. The PRD is chosen to remove the baseline or to eliminate the dc level which is added to ECG signals for storage purpose.

4.1.4 Signal to Noise Ratio

Signal-to-noise ratio (SNR) is defined as the ratio of signal power to the noise power corrupting the signal ^[32]. A ratio higher than 1:1 indicates more signal than noise.

$$\text{SNR(dB)} = 20\log_{10} \frac{\sum_{i=1}^N |x(n) - \bar{x}|^2}{\sum_{i=1}^N |e(n)|^2}$$

4.1.5 Cross Correlation Coefficients

A measure that determines the degree to which two variable's signals is associated. Value near to 1 indicates close resemblance between original and reconstructed signal.

$$\text{CC} = \frac{N(\sum xy) - (\sum x)(\sum y)}{\sqrt{[N\sum x^2 - (\sum x)^2][N\sum y^2 - (\sum y)^2]}}$$

The higher SNR, CC value and lower MAD, RMSD and PRD values means better signal restoration performance.

4.2 TVD 1D and TVD 2D ECG signal characterization

The MM approach to minimize the function $f(x)$ can be summarized as [18]:

1. Store the noisy ECG signal as data and set the number of data points.
2. Set regularization parameter λ .
3. Set number of iterations as 30.
4. Set $k = 0$. Initialize $x(0)$ as the original signal.
5. Choose $G_k(x)$ using (9) such that
 - (a) $G_k(x) > f(x)$ for all x
 - (b) $G_k(x) = f(x_k)$
6. Set $x(k+1)$ as the minimizer of $G_k(x)$.

Initially we started by taking $\lambda = 0.001$. Results were numerically good but visually poor. So, the value of λ is increased slowly and performance parameters were noted. The Table 1 shows performance matrices for different values of λ wherein bold, italic and red parameters represent result of second order TVD. From the Table 1, it is observed that at lower value of λ , performance TVD 1D is better than TVD 2D whereas at higher values, TVD 2D overtakes TVD 1D.

Table 1: Relation between first order and second order total variation denoising with respect to λ

λ $\times 10^{(-3)}$	MAD $\times 10^{(-3)}$	MAD $\times 10^{(-3)}$	RMSD $\times 10^{(-2)}$	RMSD $\times 10^{(-2)}$	PRD	PRD	SNR	SNR	CC	CC
1	2	4	12	19	0.67	1.14	50.21	45.60	1.000	0.9999
5	10	14	42	51	2.59	3.13	38.47	36.82	0.9997	0.9995
10	20	22	66	74	4.04	4.52	34.61	33.64	0.9992	0.9990
50	65	54	22	133	7.53	8.17	29.20	28.49	0.9972	0.9967
100	82	22	147	74	9.07	4.52	27.59	33.64	0.9962	0.9990

So, it can be concluded that better results can be achieved for TVD 2D at higher value of λ .

Results of both TVD 1D and TVD 2D at $\lambda = 0.01$ are indicated in Table 2 and 3 respectively.

Table 2: Quality assessment matrix of TVD 1D at $\lambda = 0.01$

Signal	MAD	RMSD	PRD	SNR	CC
100	0.0200	0.6145	3.6100	35.4131	0.9994
104	0.0200	0.5294	1.9154	36.7611	0.9998
108	0.0200	0.6568	4.0377	34.6139	0.9992
112	0.0200	0.4667	2.2984	45.8739	0.9997
115	0.0200	0.5075	1.7069	40.9574	0.9999
117	0.0200	0.5170	2.2271	44.5854	0.9998
122	0.0200	0.6214	1.7791	43.1285	0.9998
201	0.0197	0.5116	2.5891	34.4319	0.9997
205	0.0200	0.4808	2.6060	38.6971	0.9997
207	0.0200	0.6380	2.1051	35.2116	0.9998
214	0.0200	0.6309	1.3193	37.7987	0.9999
220	0.0200	0.4540	1.4573	43.3975	0.9999

Table 3: Quality assessment matrix of TVD 2D at $\lambda = 0.01$

Signal	MAD	RMSD	PRD	SNR	CC
100	0.0207	0.6755	3.9683	34.5911	0.9992
104	0.0211	0.6985	2.5274	34.3528	0.9997
108	0.0222	0.7352	4.5193	33.6351	0.9990
112	0.0171	0.4616	2.2735	45.9685	0.9997
115	0.0167	0.4879	1.6410	41.2991	0.9999
117	0.0202	0.5522	2.3786	44.0140	0.9997
122	0.0248	0.6920	1.9812	42.1940	0.9998
201	0.0175	0.6027	3.0501	33.0086	0.9995
205	0.0181	0.4245	2.3008	39.7792	0.9997
207	0.0233	0.7586	2.5033	33.7068	0.9997
214	0.0212	0.7645	1.5988	36.1302	0.9999
220	0.0198	0.4778	1.5334	42.9549	0.9999

More or less the performance parameters of TVD 1D and TVD 2D are same. However, second difference is more suitable as staircase effect is

significantly reduced which can be seen in the Figures 2-3.

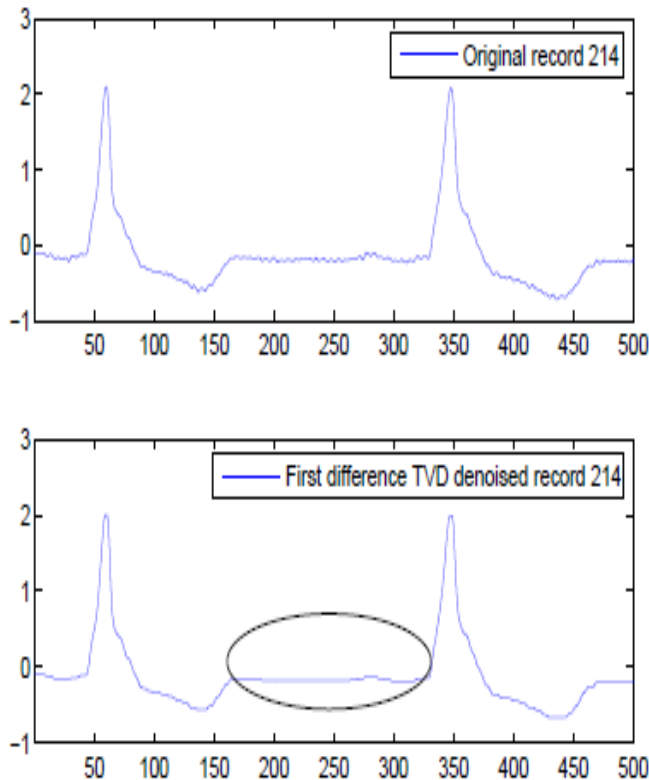


Figure 2: Original and TVD 1D reconstructed record number 214

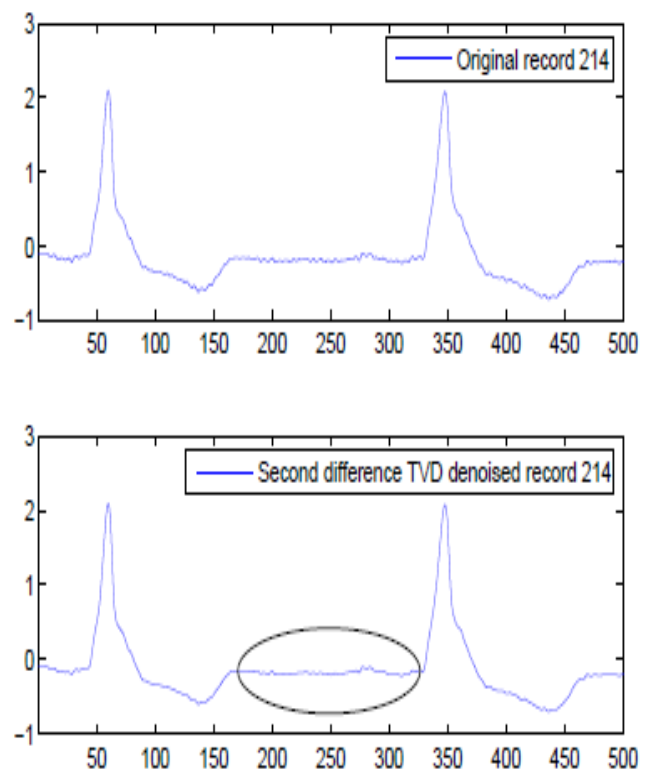


Figure 3: Original and TVD 2D reconstructed record number 214

4.3 ECG characterization through Lagrange-Chebyshev interpolation

Here we need to construct an interpolating polynomial $p_n(x)$ using (24) with the N ECG samples using the Chebyshev nodes. Since an ECG signal sampled value may not be available at all the Chebyshev nodes, we derive these values by linear interpolation using adjacent ECG sampled values. We continue increasing the order till our error criterion is met. This approximation technique can be summarized as:

1. Fix the order n and the tolerance $\xi = 10^{-4}$ for the Lagrange-Chebyshev polynomial approximation.
2. Transform the Chebyshev nodes on the domain $[a, b]$ and calculate the zeros i.e., the Chebyshev nodes x_j using (22).
3. Find the function value $f(x_j)$ by linear interpolation using the two adjacent samples around x_j .
4. Construct interpolating polynomial $p_n(x)$ using (24).
5. Calculate error $E_n(x) = \max |f(x) - p_n(x)|$.

6. If $E_n(x) > \xi$ then $n = n + 1$ and go to step 2.

In order to denoise the whole signals, high order polynomials were required which increases computation time. Also errors for this approach were very high. So, to reduce the computation time and error, we segmented the whole signal into suitable (till optimum results are achieved) number of segments. This is done by implementing Bottom up approach. The Bottom up algorithm, also called as iterative merge which begins by dividing the original time series data of length n , into a large number of segments and is consequently merged into bigger segments until stopping criteria is met [33]. So, segmentation is done before interpolation. The signal reconstruction stage consists of sequentially appending the segments to obtain the complete reconstructed signal which does not require any selection of significant coefficients. Here the 10 seconds (3600 samples) of original signal from MIT-BIH database is divided into 100 segments and then the individual segments are interpolated by 50th order Lagrange-Chebyshev interpolants. Table 4 shows the results of Lagrange-Chebyshev interpolation for same set of signals.

Table 4: Quality assessment matrices for Lagrange-Chebyshev interpolation

Signal	MAD $\times 10^{-6}$	RMSD $\times 10^{-5}$	PRD	SNR	CC
100	1.83	6.16	0.0019	44.83	0.9977
104	2.89	9.55	0.0017	43.06	0.9982
108	0.15	1.12	1.7153×10^{-4}	66.91	1.0000
112	0.80	2.87	4.6281×10^{-4}	55.97	0.9999
115	1.81	5.25	0.0030	50.34	0.9945
117	2.59	13.00	0.0014	43.65	0.9987
122	0.39	11.32	0.0023	47.61	0.9968
201	0.84	3.39	9.7624×10^{-4}	44.81	0.9994
205	0.70	3.83	1.1696×10^{-4}	53.05	1.0000
207	12.01	24.68	0.0139	23.17	0.8837
214	5.70	16.91	0.0090	32.29	0.9535
220	3.93	14.48	0.0031	41.46	0.9941

Although the number of segments has been kept same for all the signals, all the segments cannot be approximated by a single polynomial because they are of unequal sample lengths due to the variable shape of ECG within and across patients. Here our interest is to denoise the signal

while retaining its characteristics. So, number of segments is not so important.

The performance matrices for the total variation are more or less similar. RMSD value is of order 10^{-1} whereas MAD values are upto 10^{-2} . Maximum value of PRD is 4 which can be

considered to be within diagnostic limits. SNR and CC values are also high indicating high level of noise suppression and high resemblance of timing pattern of reconstructed signal with the original signal. So, both the TVD 1D and TVD 2D methods are suitable for noise suppression of ECG signals. However, visually better results are obtained for TVD 2D. As compared to these techniques, results of Lagrange- Chebyshev are much better. Errors are significantly reduced up to 10^{-4} . Signal contents

are also enhanced to high level as compared to noise level. CC values are found to be less than the methods which can be improved either by increasing number of segments or by increasing order of the polynomial. Original and Lagrange-Chebyshev approximated record 214 is presented in Figure 4. Thus the proposed method will be more suitable for ECG denoising.

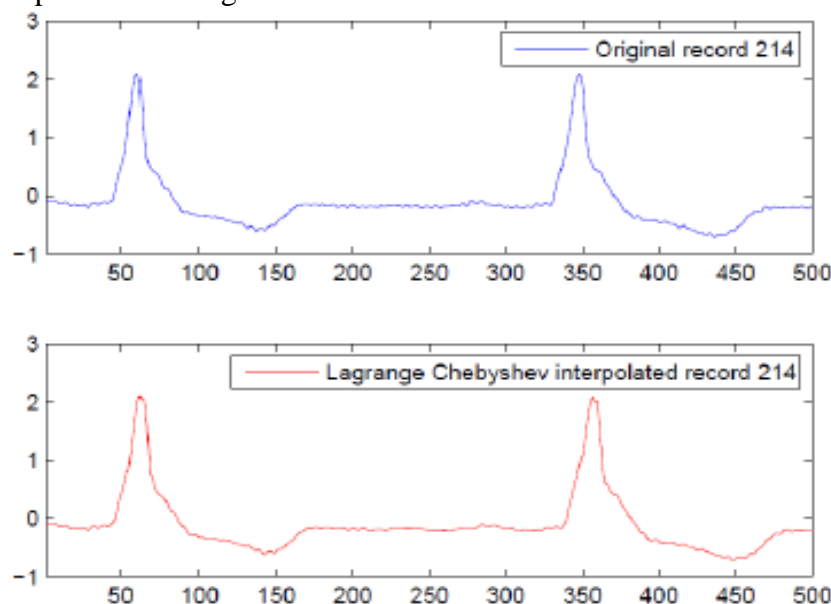


Figure 5: Original and Lagrange –Chebyshev reconstructed record number 214

5. Conclusion

ECG signal records voltage potentials of cardiac activity. These signals are often contaminated by various types of noises. In this paper, ECG signals from MIT-BIH database were denoised using TVD 1D, 2D and Lagrange-Chebyshev interpolation. It is concluded that both methods of TVD are suitable for ECG denoising but TVD 2D results are visually better. Also regularization parameter which controls degree of denoising plays an important role. Same set of signals are denoised using Lagrange- Chebyshev interpolation method of high order and results obtained are comparable. Results are further improved by segmentating ECG signals using Bottom up approach. The results obtained are diagnostically acceptable and are found to be superior to those reported in the existing literature.

Moreover, it was observed that the segmentation helps in considerable reduction of computation time.

The accuracy can be further improved by breaking the complete signal into more number of segments but that could be at the cost of CR and computation time. A combination of TVD 1D-2D may provide better results.

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